

# Medical Image Processing Using Transforms

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Fields, 08, HmZhu

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## Outlines

- Image Quality
- Gray value transforms
- Histogram processing
- Filters in image space
- **Filters in Fourier space**
- Filters in Time-frequency space

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## 5. Filters in Fourier space

5.1. The Fourier transforms



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- In 1807, his introductory manuscript "Theory of the propagation of heat in solid bodies"
- In 1822, published "Theorie analytique de la chaleur"
- Established the mathematical theory of heat diffusion
- Introduced the representation of a function as a series of sines and cosines known as Fourier series

$$u_t = ku_{xx}$$

Elena Prestini: evolution of applied harmonic analysis (2004)

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### The Fourier Transform (1807)

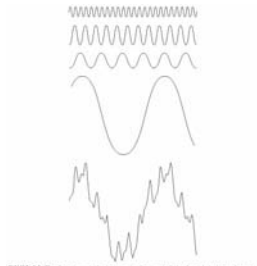
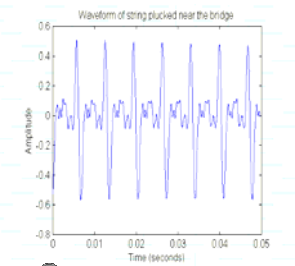


FIGURE 4.1 The function at the bottom is the sum of the four functions above it. Fourier's idea in 1807 that periodic functions could be represented as a weighted sum of sines and cosines was met with skepticism.

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### The Fourier Transform

$$f(x) \xrightarrow{\text{Fourier transform}} F(u) = \int_{-\infty}^{+\infty} f(x) \exp(-j2\pi ux) dx$$

$$f(x) = \int_{-\infty}^{+\infty} F(u) \exp(j2\pi ux) du \xleftarrow{\text{Inverse Fourier Transform}} F(u)$$



Joseph Fourier

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# The Fourier Transform

The FT of a function  $f(t)$  is in general complex:

$$F(u) = Re(u) + j Im(u)$$

Fourier spectrum:  $|F(u)| = \sqrt{Re^2(u) + Im^2(u)}$

Power spectrum (or spectral density):  $|F(u)|^2$

Phase angle:  $\phi(u) = \tan^{-1} \left( \frac{Im(u)}{Re(u)} \right)$

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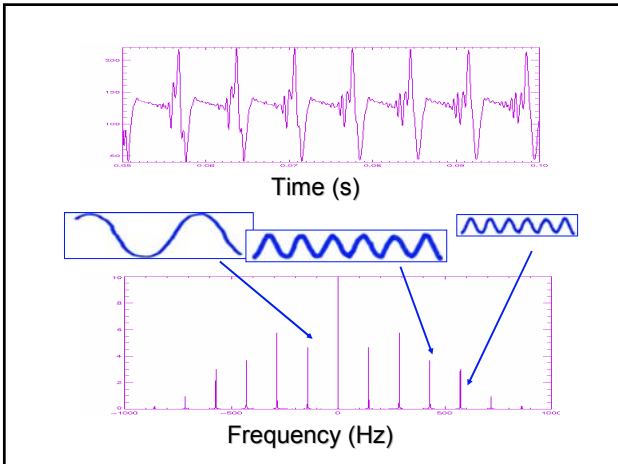
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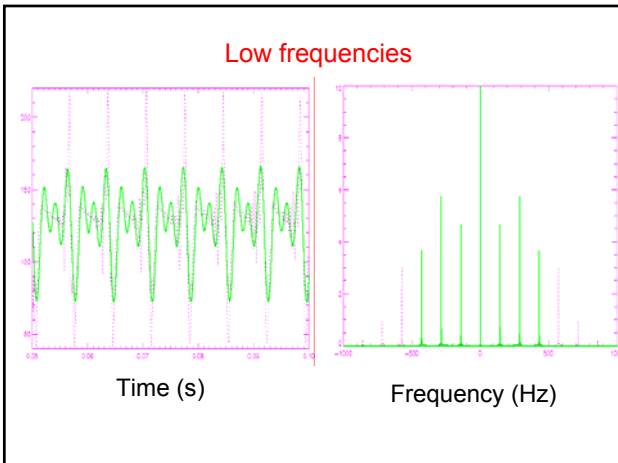
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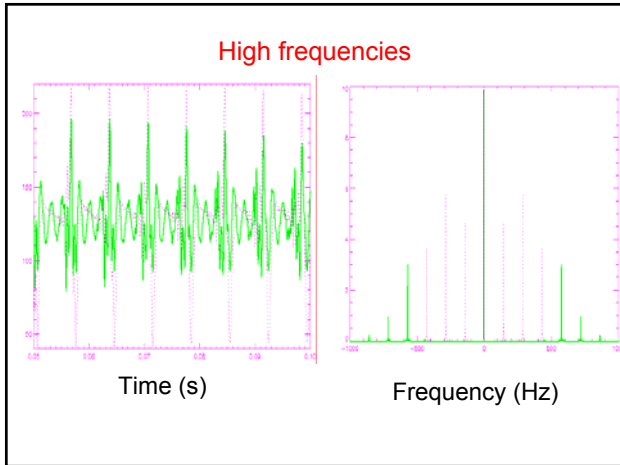
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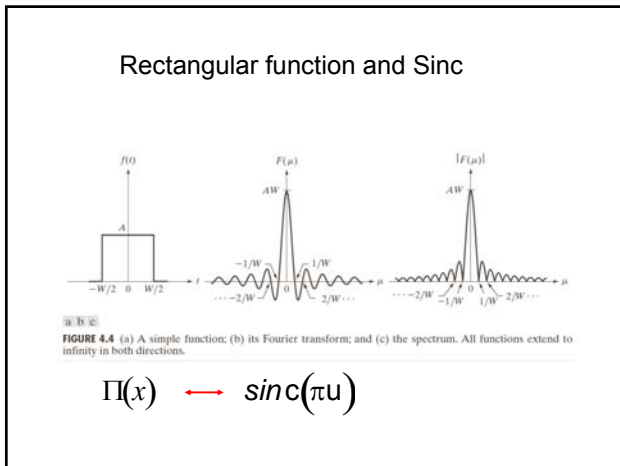
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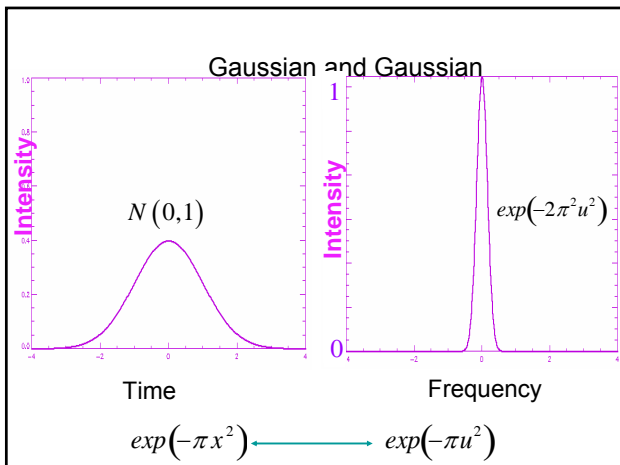
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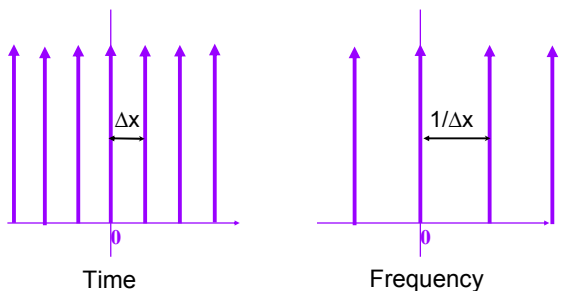
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### Comb and Comb



$$\frac{1}{\Delta x} \operatorname{comb}\left(\frac{x}{\Delta x}\right) \longleftrightarrow \operatorname{comb}\left(\frac{u}{\Delta u}\right)$$

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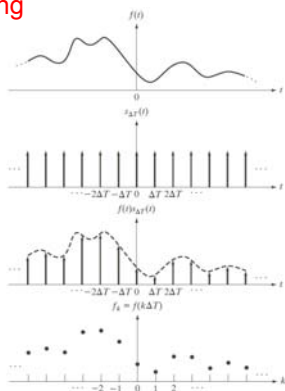
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### Sampling



**FIGURE 4.5** (a) A continuous function. (b) Train of impulses used to model the sampling process. (c) Sampled function formed as the product of (a) and (b). (d) Sample values obtained by integration and using the shifting property of the impulse. (The dashed line in (c) is shown for reference; it is not part of the data.)

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### Some Fourier Properties

Scaling:  $h(ax) \longleftrightarrow \frac{1}{|a|} H\left(\frac{u}{a}\right)$

Shifting:  $h(x - x_0) \longleftrightarrow H(u)e^{-j2\pi ux_0}$

$h(x)e^{j2\pi u_0 x} \longleftrightarrow H(u - u_0)$

Convolution:  $h_1(x) \otimes h_2(x) \longleftrightarrow H_1(u)H_2(u)$

$h_1(x)h_2(x) \longleftrightarrow H_1(u) \otimes H_2(u)$

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## Convolution Theorem

Definition - Convolution:

$$h_1(x) \otimes h_2(x) \equiv \int_{-\infty}^{+\infty} h_1(\tau) h_2(x - \tau) d\tau$$

Property:

$$h_1 \otimes h_2 = h_2 \otimes h_1$$

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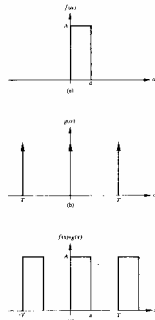
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## Convolution with impulse function




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## Convolution = Blurring

Convoluting a signal with a smooth weighting function can be used to smooth a signal

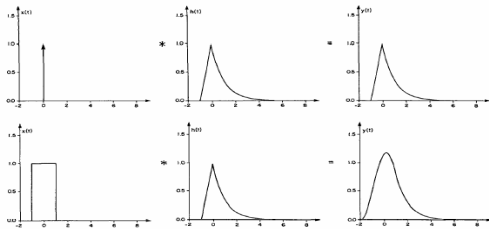


Fig. 3.8: The results of convoluting an impulse response with an impulse (top) and a square pulse (bottom) are shown here.

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What happens in 2D?

The 2D Fourier Transform



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The 2D Fourier Transform

$$F(k_x, k_y) = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} f(x, y) e^{-j2\pi k_x x} dx \cdot e^{-j2\pi k_y y} dy$$

↑

$$f(x, y) = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} F(k_x, k_y) e^{j2\pi k_x x} dk_x e^{j2\pi k_y y} dk_y$$

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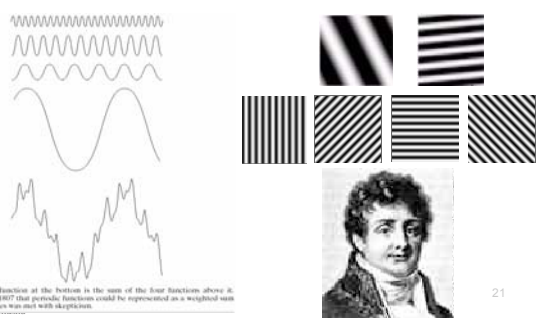
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Review of Fourier Theory



**FIGURE 4.1** The function at the bottom is the sum of the four functions above it. Fourier's idea in 1807 that periodic functions could be represented as a weighted sum of sines and cosines was met with skepticism.

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What are Spatial Frequencies?

Image

Fourier Spectrum

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What's wrong?

What happened to the image?

Incorrect Fourier spectra

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Low Spatial Frequencies

Basic structure of an Image

Low frequencies

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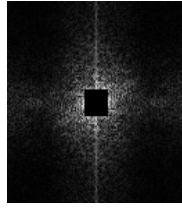


## High Spatial Frequencies



Edges of  
an image

FT  
↔



High frequencies

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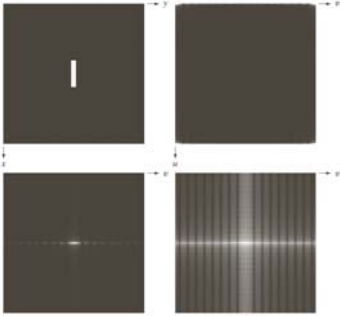
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## Amplitude



**FIGURE 4.24**  
(a) Image.  
(b) Spectrum showing bright spots in the four corners.  
(c) Central spectrum.  
(d) Result showing increased detail after a log transformation. The axes crossings of the spectrum are closer in the vertical direction because the rectangle in (a) is longer in that direction. The coordinate conventions used throughout the book place the origin of the spatial and frequency domains at the top left.

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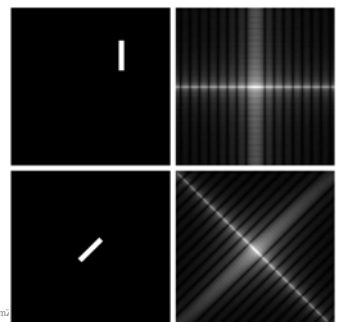
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## Amplitude



**FIGURE 4.25**  
(a) The rectangle in Fig. 4.24(a) translated, and (b) the corresponding spectrum.  
(c) Rotated rectangle, and (d) the corresponding spectrum. The spectrum corresponding to the translated rectangle is identical to the spectrum corresponding to the original image in Fig. 4.24(a).

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Phase

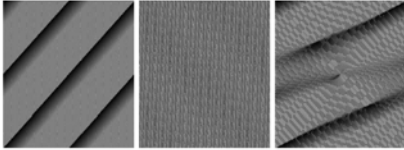


FIGURE 4.26 Phase angle array corresponding (a) to the image of the centered rectangle in Fig. 4.24(a), (b) to the translated image in Fig. 4.25(a), and (c) to the rotated image in Fig. 4.25(c).

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Amplitude and Phase



FIGURE 4.27 (a) Woman. (b) Phase angle. (c) Woman reconstructed using only the phase angle. (d) Woman reconstructed using only the spectrum. (e) Reconstruction using the phase angle corresponding to the woman and the spectrum corresponding to the rectangle in Fig. 4.24(a). (f) Reconstruction using the phase of the rectangle and the spectrum of the woman.

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Basic properties still hold

Name	DFT Pairs
1) Symmetry properties	See Table 4.1
2) Linearity	$af_1(x, y) + bf_2(x, y) \Leftrightarrow aF_1(u, v) + bF_2(u, v)$
3) Translation (general)	$f(x, y)e^{j2\pi(ux_0 + vy_0)} \Leftrightarrow F(u - u_0, v - v_0)$ $f(x - x_0, y - y_0) \Leftrightarrow F(u, v)e^{j2\pi(ux_0 + vy_0)}$
4) Translation to center of the frequency rectangle, $(M/2, N/2)$	$f(x, y)(-1)^{x+y} \Leftrightarrow F(u - M/2, v - N/2)$ $f(x - M/2, y - N/2) \Leftrightarrow F(u, v)(-1)^{x+y}$
5) Rotation	$f(r, \theta + \theta_0) \Leftrightarrow F(u, v + \theta_0)$ $x = r \cos \theta \quad y = r \sin \theta \quad u = \omega \cos \varphi \quad v = \omega \sin \varphi$
6) Convolution theorem <sup>1</sup>	$f(x, y) \star h(x, y) \Leftrightarrow F(u, v)H(u, v)$ $f(x, y)h(x, y) \Leftrightarrow F(u, v) \star H(u, v)$

TABLE 4.3 Summary of DFT pairs. The closed-form expressions in 12 and 13 are valid only for continuous variables. They can be used with discrete variables by sampling the closed-form continuous expressions.

(Continued)

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## Basic properties still hold

Name	DFT Pairs
7) Correlation theorem'	$f(x, y) \otimes h(x, y) \Rightarrow F^*(u, v)H(u, v)$ $f^*(x, y)h(x, y) \Rightarrow F(u, v)H(u, v)$
8) Discrete unit impulse	$\delta(x, y) \Rightarrow 1$
9) Rectangle	$\text{rect}[u, b] \Rightarrow ab \frac{\sin(\pi ub)}{(\pi ub)} \frac{\sin(\pi vb)}{(\pi vb)} e^{-j2\pi(ub)}$
10) Sine	$\sin(2\pi u_x + 2\pi v_y) \Rightarrow \frac{j}{2}[\delta(u + M u_0, v + N v_0) - \delta(u - M u_0, v - N v_0)]$
11) Cosine	$\cos(2\pi u_x + 2\pi v_y) \Rightarrow \frac{1}{2}[\delta(u + M u_0, v + N v_0) + \delta(u - M u_0, v - N v_0)]$
The following Fourier transform pairs are derivable only for continuous variables, denoted as before by $r$ and $z$ for spatial variables and by $\mu$ and $\nu$ for frequency variables. These results can be used for DFT work by sampling the continuous forms.	
12) Differentiation	$\left(\frac{\partial}{\partial r}\right) \left(\frac{\partial}{\partial z}\right) f(r, z) \Rightarrow (j2\pi\mu)^r (j2\pi\nu)^z F(\mu, \nu)$ (The expressions on the right assume that $\frac{\partial^2 f(r, z)}{\partial r^2} \Rightarrow (j2\pi\mu)^2 F(\mu, \nu)$ , $\frac{\partial^2 f(r, z)}{\partial z^2} \Rightarrow (j2\pi\nu)^2 F(\mu, \nu)$ , $f(r=0, z=0) = 0$ .)
13) Gaussian	$A \exp(-\alpha^2 r^2 - \beta^2 z^2) \Rightarrow A \exp(-j^2 \alpha^2 \mu^2 - \beta^2 \nu^2)$ ( $A$ is a constant)

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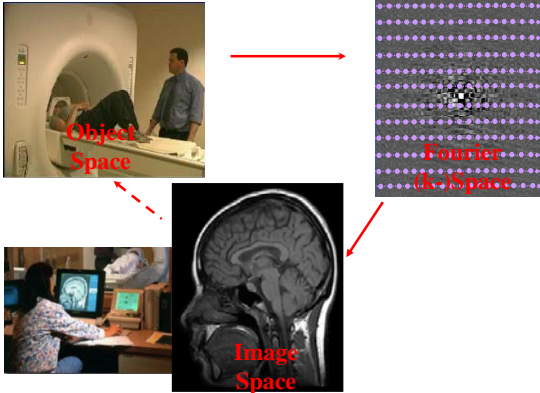
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## MR Image Reconstruction




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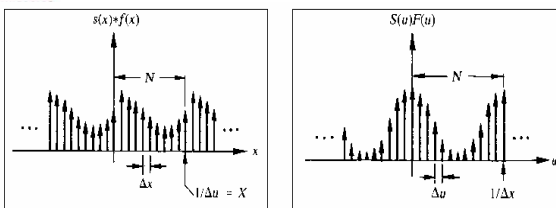
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## The 1D Discrete FT



Object space and Fourier space is kind of "inverse" relationship

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## The 2D Discrete FT

Field of View (FOV)

K extend

$$\Delta u = \frac{1}{X} \quad \Delta v = \frac{1}{Y}$$

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### Sampling

aliasing

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## Example: Spatial Aliasing

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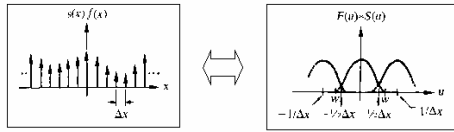
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### Nyquist Sampling Criterion



To avoid aliasing,  $\frac{1}{2\Delta x} \geq w$  i.e.  $\Delta x \leq \frac{1}{2w}$

w: Nyquist frequency                      2w: Nyquist rate

That is, the sampling rate needs to be at least twice the max frequency of the function:  
 $\frac{1}{\Delta x} \geq 2w$

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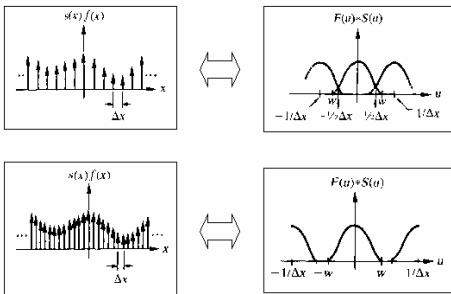
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### Infinite Sampling



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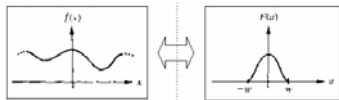
### How much data is needed ?

#### Shannon Sampling Theorem

If  $h(t)$  is a band-limited continuous function,  
 i.e. there is a finite positive  $w$ , such that

$$H(u) = 0 \text{ for } |u| > w$$

then  $h(t)$  can completely recover from samples whose spacing satisfies Nyquist criterion



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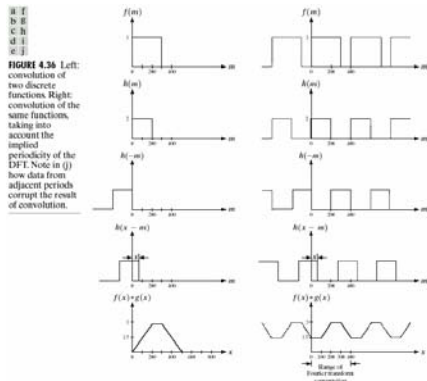
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**Warning:** Using DFT allows us to perform convolution, but the discrete functions are treated as periodic, with a periodic equal to the length of the functions. Wrap-around error could happen.




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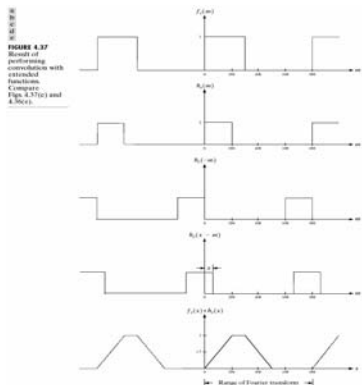
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**Solution:** Assume that  $f$  and  $h$  consist of  $A$  and  $B$  points, respectively. We append zeros to both so that they have the identical periods  $P \geq A+B - 1$ .




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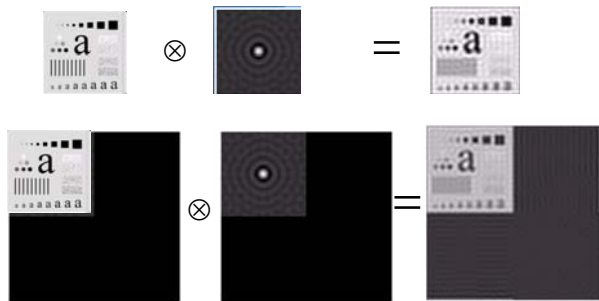
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**Solution:** We zero-extends  $f(x, y)$  and  $h(x, y)$  so that both have size of  $P \times Q$  where  $P \geq A+C - 1$  and  $Q \geq B+D-1$ .




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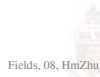
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# 5. Filters in Fourier space

## 5.2. Filters in Fourier space




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$H(u, v)$  is called filter or filter transfer function, which suppresses certain frequencies while leaving the others unchanged.

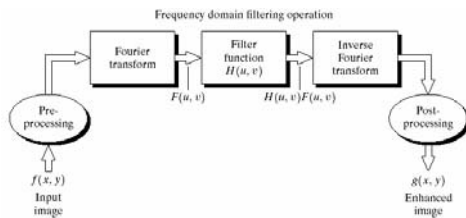


FIGURE 4.5 Basic steps for filtering in the frequency domain.

$$f(x, y) \otimes h(x, y) \Leftrightarrow F(u, v)H(u, v)$$

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## Example

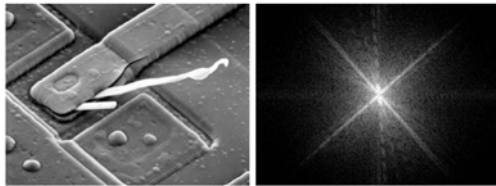


FIGURE 4.29 (a) SEM image of a damaged integrated circuit. (b) Fourier spectrum of (a). (Original image courtesy of Dr. J. M. Hudak, Brockhouse Institute for Materials Research, McMaster University, Hamilton, Ontario, Canada.)

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**Low pass filters:**  
attenuates high freq. while passing low freq (smoothing, blurring)

**High pass filters:**  
attenuates low freq while passing high freq (sharpening)

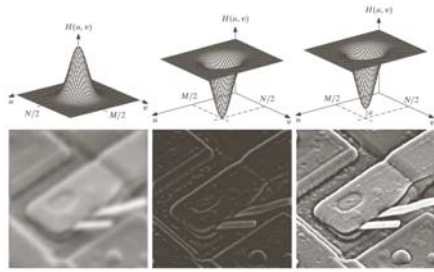


FIGURE 4.31 Top row: frequency domain filters. Bottom row: corresponding filtered images obtained using Eq. (4.7-1). We used  $\alpha = 0.85$  in (c) to obtain (f) (the height of the filter itself is 1). Compare (f) with Fig. 4.29(a).

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### Low pass filters

**TABLE 4.4**  
Lowpass filters.  $D_0$  is the cutoff frequency and  $n$  is the order of the Butterworth filter.

Ideal	Butterworth	Gaussian
$H(u, v) = \begin{cases} 1 & \text{if } D(u, v) \leq D_0 \\ 0 & \text{if } D(u, v) > D_0 \end{cases}$	$H(u, v) = \frac{1}{1 + [D(u, v)/D_0]^{2n}}$	$H(u, v) = e^{-D(u, v)/2D_0}$

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### Low pass filters

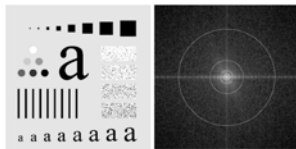


FIGURE 4.41 (a) Test pattern of size  $688 \times 688$  pixels, and (b) its Fourier spectrum. The spectrum is double the image size due to packing but is shown in half size so that it fits in the page. The superimposed circles have radii equal to 0, 20, 60, 100, and 400 with respect to the full-size spectrum image. These radii enclose 87.8, 93.1, 95.7, 97.8, and 99.2% of the packed image power, respectively.

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Ideal Low Pass Filter



FIGURE 4.42 (a) Perspective plot of an ideal low-pass filter transfer function. (b) Filter displayed as an image. (c) Filter radial cross section.

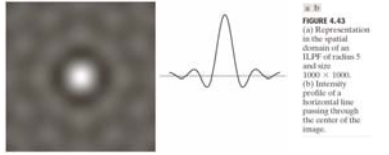


FIGURE 4.43 (a) Representation of an LLPF of radius 5 and size 1000 x 1000. (b) Intensity profile of a horizontal line passing through the center of the image.

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Ideal low pass filter



FIGURE 4.43 (a) Original image. (b)-(i) Results of filtering using LLPFs with small variations of radii. (b) 5, (c) 10, (d) 15, (e) 20, (f) 25, (g) 30, (h) 35, (i) 40. The images are arranged in the order (a), (b), (c), (d), (e), (f), (g), (h), (i).

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Butterworth low pass filters

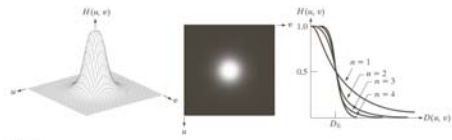


FIGURE 4.44 (a) Perspective plot of a Butterworth low-pass filter transfer function. (b) Filter displayed as an image. (c) Filter radial cross sections of orders 1 through 4.



FIGURE 4.44 (a)-(d) Spatial representations of Butterworth low-pass filters of orders 1, 2, 3, and 4, and corresponding intensity profiles through the center of the filters (the size is all cases is 1000 x 1000 and the cutoff frequency is 5). Observe how ringing increases as a function of filter order.

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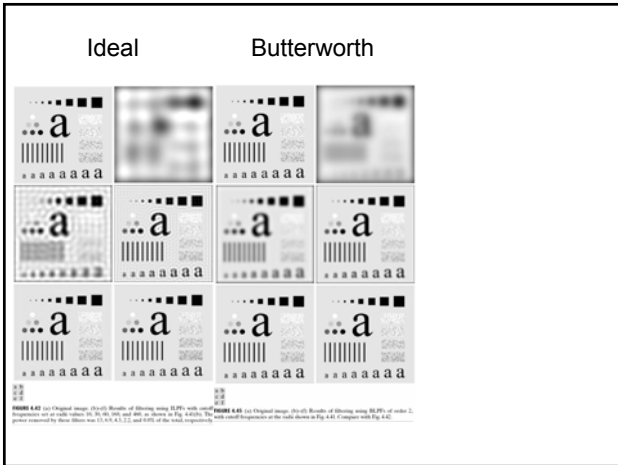
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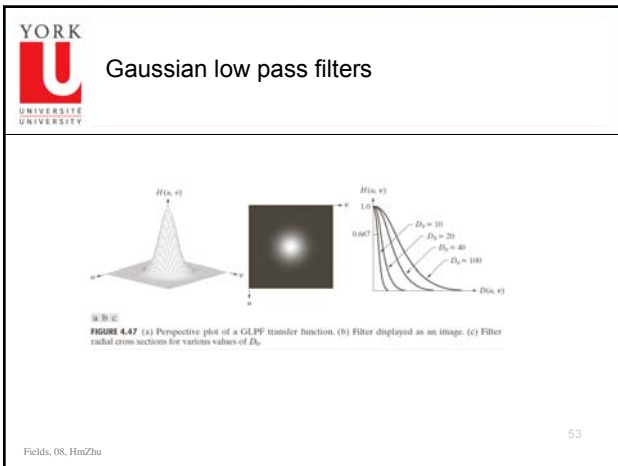
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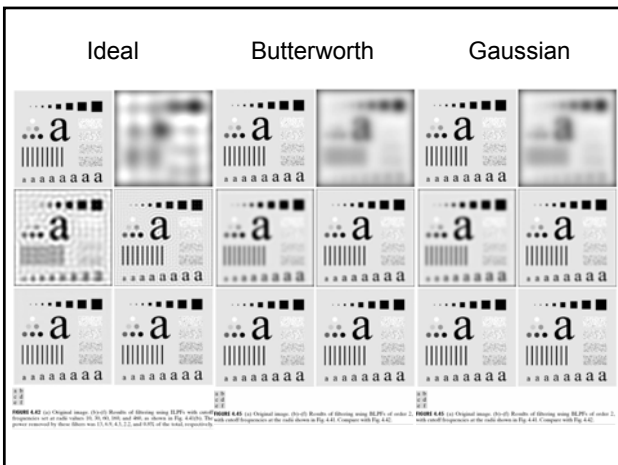
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## High pass filters

$$H_{HP}(u, v) = 1 - H_{LP}(u, v)$$

**TABLE 4.5**  
Highpass filters.  $D_0$  is the cutoff frequency and  $n$  is the order of the Butterworth filter.

Ideal	Butterworth	Gaussian
$H(u, v) = \begin{cases} 0 & \text{if } D(u, v) \leq D_0 \\ 1 & \text{if } D(u, v) > D_0 \end{cases}$	$H(u, v) = \frac{1}{1 + [D_0/D(u, v)]^{2n}}$	$H(u, v) = 1 - e^{-D^2(u, v)/2D_0^2}$

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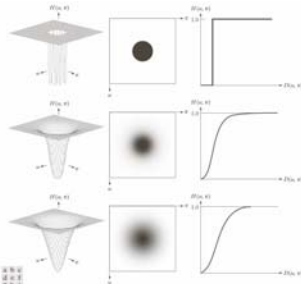
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## High pass filters



**FIGURE 4.52** Top row: Perspective plots, image representations, and cross sections of a typical ideal highpass filter. Middle and bottom rows: The same responses for critical Butterworth and Gaussian highpass filters.

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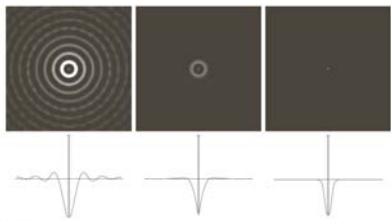
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## Spatial representations of the high pass filters



**FIGURE 4.53** Spatial representation of typical (a) ideal, (b) Butterworth, and (c) Gaussian frequency domain highpass filters, and corresponding intensity profiles through their centers.

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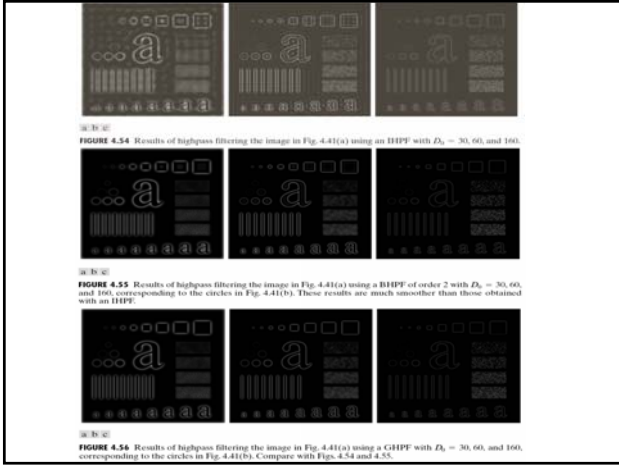
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### An example of high pass filter

**FIGURE 4.57** (a) Thumb print. (b) Result of highpass filtering (a). (c) Result of thresholding (b). (Original image courtesy of the U.S. National Institute of Standards and Technology.)

Fields, 08, HmZhu

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### Unshark masking and High-frequency emphasis filter

The mask in the unsharp masking can be defined as

$$g_{mask}(x, y) = f(x, y) - f_{LP}(x, y)$$

where  $f_{LP}(x, y) = \mathbb{F}^{-1} \{ [H_{LP}(u, v) F(u, v)] \}$ .

Then the final filtered image can be expressed as

$$g(x, y) = f(x, y) + k * g_{mask}(x, y)$$

$$= \mathbb{F}^{-1} \{ [ [1 + k * H_{HP}(u, v)] F(u, v) ] \}$$

More general formulation of the high-frequency emphasis filtering

$$g(x, y) = \mathbb{F}^{-1} \{ [ [k_1 + k_2 * H_{HP}(u, v)] F(u, v) ] \}$$

where  $k_1, k_2 \geq 0$ .

Fields, 08, HmZhu

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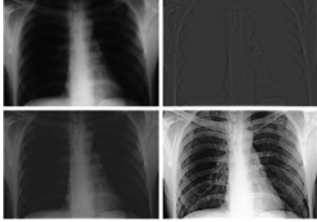
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## High-frequency emphasis filter



**FIGURE 4.59** (a) A chest X-ray image. (b) Result of highpass filtering with a Gaussian filter. (c) Result of high-frequency-emphasis filtering using the same filter. (d) Result of performing histogram equalization on (c). (Original image courtesy of Dr. Thomas R. Gest, Division of Anatomical Sciences, University of Michigan Medical School.)

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## Bandreject filters

**TABLE 4.6**  
Bandreject filters.  $W$  is the width of the band.  $D$  is the distance  $D(u, v)$  from the center of the filter.  $D_0$  is the cutoff frequency, and  $n$  is the order of the Butterworth filter. We show  $D$  instead of  $D(u, v)$  to simplify the notation in the table.

Ideal	Butterworth	Gaussian
$H(u, v) = \begin{cases} 0 & \text{if } D_0 - \frac{W}{2} \leq D \leq D_0 + \frac{W}{2} \\ 1 & \text{otherwise} \end{cases}$	$H(u, v) = \frac{1}{1 + \left[ \frac{DW}{D_0^2 - D^2} \right]^{2n}}$	$H(u, v) = 1 - e^{-\left( \frac{DW}{D_0^2} \right)^2}$

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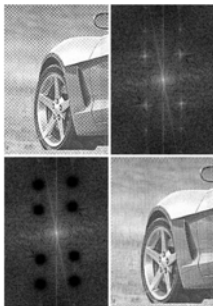
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## An Example of Bandreject Filter



**FIGURE 4.64**  
(a) Sampled newspaper image showing a motor pattern. (b) Spectrum. (c) Butterworth notch reject filter multiplied by the Fourier transform. (d) Filtered image.

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# 5. Filters in Fourier space

## 5.3. Deconvolution

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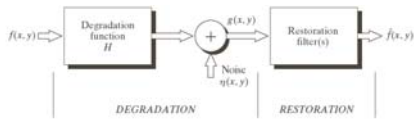
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# Image restoration and reconstruction

FIGURE 5.1 A model of the image degradation/restoration process.



The measured image can be considered as

$$g(x, y) = h(x, y) \otimes f(x, y) + n(x, y)$$

$$G(u, v) = H(u, v)F(u, v) + N(u, v)$$

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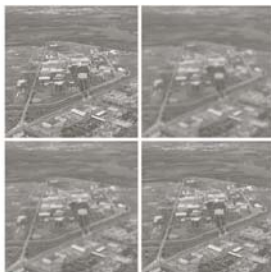
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# Degraded images

FIGURE 5.25 Illustration of the atmospheric turbulence model. (a) Negligible turbulence,  $\lambda = 0.43 \mu\text{m}$ . (b) Medium turbulence,  $\lambda = 0.63 \mu\text{m}$ . (c) Low turbulence,  $\lambda = 0.83 \mu\text{m}$ . (Original image courtesy of NASA.)




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## Inverse filter

The measured image can be considered as

$$g(x, y) = h(x, y) \otimes f(x, y) + n(x, y)$$

↓ FT

$$G(u, v) = H(u, v)F(u, v) + N(u, v)$$

Given the Fourier response of the degrading function  $H(u, v)$ , the true image can be estimated by

$$\tilde{F}(u, v) = \frac{G(u, v)}{H(u, v)} = F(u, v) + \frac{N(u, v)}{H(u, v)}$$

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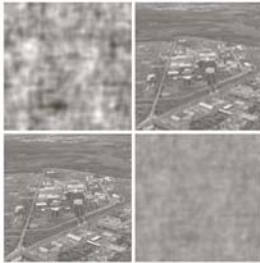
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## Inverse filter

FIGURE 5.27  
Blurring  
Fig. 5.27(a) with  
Fig. 5.27(b)  
(c) Result of  
using the  
filter (d) Result  
with a radius of  
40 (e) result of  
radius of 70 and  
(f) result of  
radius of 95.




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## Wiener filter (1942)

The Wiener filter aims to find an estimate to the true image by minimizing their mean square error.

In the Fourier domain, the spectrum of the estimated image is

$$\tilde{F}(u, v) = \frac{H^*(u, v)}{\left[ |H(u, v)|^2 + \frac{|N(u, v)|^2}{|F(u, v)|^2} \right]} G(u, v)$$

$$\approx \frac{1}{\left[ |H(u, v)|^2 + K \right]} G(u, v)$$

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## Wiener filter



FIGURE 5.28 Comparison of inverse and Wiener filtering. (a) Result of full inverse filtering of Fig. 5.25(b). (b) Radially limited inverse filter result. (c) Wiener filter result.

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## Wiener filter



FIGURE 5.29 (a) 8-bit image corrupted by motion blur and additive noise. (b) Result of inverse filtering. (c) Result of Wiener filtering. (d)–(f) Same sequence, but with noise variance one order of magnitude less. (g)–(i) Same sequence, but noise variance reduced by five orders of magnitude from (a). Note in (h) how the deblurred image is quite visible through a “curtain” of noise.

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## 5. Filters in Fourier space

### 5.4. Applications

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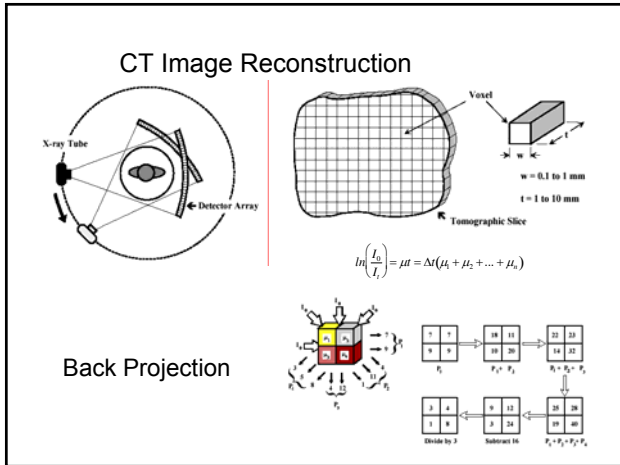
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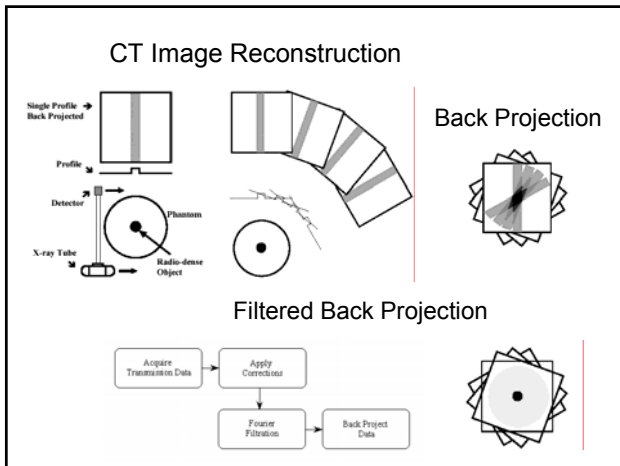
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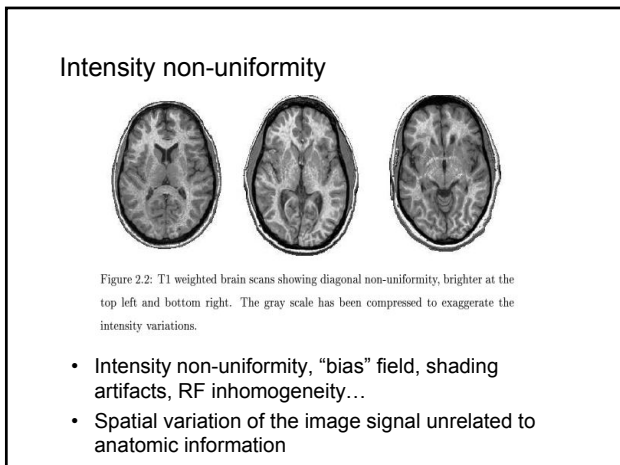
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## Causes and effect

### Major factors:

- Non-uniform reception sensitivity (multiplicative field, slow varying)
- Inhomogeneity of RF excitation (non-multiplicative field; acquisition)
- RF penetration and standing wave effects (image acquisition)

### Minor factors:

- eddy currents driven by the switching of field gradients
  - Mis-tuning of the RF coil
  - Bandwidth filtering of the data
  - Geometric distortion (negligible, 1%)
- Small; The effect increases with the rapid image acquisition
- Automatically done by scanners

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## Effects

- Cause 10% - 30% variation of intensity.
- Reduce accuracy of tissue segmentation, brain-surface extraction
- The higher the field strength, the prominent the intensity non-uniformity

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## Non-uniformity correction Methods (N3 Method)

### A Nonparametric Method for Automatic Correction of Intensity Nonuniformity in MRI Data

John G. Sled,\* Alex P. Zijdenbos, Member, IEEE, and Alan C. Evans

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## Assumptions and Tasks

### Assumption

- Bias field varies smoothly across images
- Model of intensity non-uniformity as a multiplicative field corrupting the measured intensities

### Tasks

- To estimate the bias field and true tissue intensity distribution
- To remove the bias field

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## Overview of Methods

Analytic modeling (too simple, impractical)  
Adjust acquisition protocols

Data-driven post-processing:

use filter to estimate a multiplicative non-uniformity field & divide it from the image

eliminate the low frequency components of anatomy when estimating non-uniformity (tissue intensity, spatial homogeneity, field based models)

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## Overview of the N3 method

### Nonuniformity Model

Smooth multiplicative field

$$v(\mathbf{x}) = u(\mathbf{x})f(\mathbf{x}) + n(\mathbf{x})$$

measured  $v(\mathbf{x})$     true?  $f(\mathbf{x})$     bias field?  $u(\mathbf{x})$     white noise?  $n(\mathbf{x})$

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Noise issue

$$v(\mathbf{x}) = u(\mathbf{x})f(\mathbf{x}) + n(\mathbf{x})$$

Filtering

$$v(\mathbf{x}) = u(\mathbf{x})f(\mathbf{x})$$

Correction

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Overview of the N3 method

$$\hat{v}(\mathbf{x}) = \hat{u}(\mathbf{x}) + \hat{f}(\mathbf{x}) \quad \hat{u}(\mathbf{x}) = \log(u(\mathbf{x}))$$

Let  $U$ ,  $V$ , and  $F$  be the probability densities of  $\hat{u}$ ,  $\hat{v}$ , and  $\hat{f}$  respectively.

$$V(\hat{v}) = F(\hat{f}) * U(\hat{v})$$

Non-uniformity distribution  $F$  blurs the histogram of the data  $U$

The task is to restore  $U$  by sharpening  $V$

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Estimating  $F$

$$V(\hat{v}) = F(\hat{f}) * U(\hat{v})$$

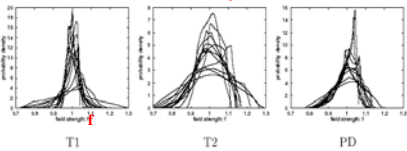


Figure 4.1: Probability densities for non-uniformity fields estimated from white matter on MR scans of twelve individuals taken with twelve different MR scanners. Note that these are the distributions of the dimensionless scale factors  $f$  rather than the log of these values  $\hat{f}$ . The T1 scan is a gradient echo 3D acquisition with TE = 11 ms and TR = 35 ms, while the proton density (PD) and T2 scans are two echoes TE = 30 ms and TE = 80 ms of a multi-slice spin echo acquisition with TR = 3s.

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### Estimating F

$$V(\hat{v}) = F(\hat{v}) * U(\hat{v})$$

- Assume that the distribution of F is Gaussian
- Searching an optimal U by finding a Gaussian distributed F having zero mean and given variance
- In other word, the space of all distributions U is collapsed down to a single dimension, the width of the F distribution

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### Estimating F

$$V(\hat{v}) = F(\hat{v}) * U(\hat{v})$$

- The Gaussian distributed F can be estimated incrementally by convolving narrow Gaussians
- Thus, U can be searched iteratively by deconvolving narrow Gaussians

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### Estimating U

$$\begin{aligned}
 V(\hat{v}) &= F(\hat{v}) \otimes U(\hat{v}) \\
 &\quad \downarrow \text{FT} \\
 \tilde{V}(k) &= \tilde{F}(k) \bullet \tilde{U}(k) \\
 &\quad \downarrow \tilde{F}^* \\
 \tilde{U} &= \frac{\tilde{F}^*}{|\tilde{F}|^2 + Z^2} \tilde{V}
 \end{aligned}$$

Constant term to avoid zero division

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## Estimating the log field $\hat{f}_e$

- Given the distributions  $F$  and  $U$ , estimate the corresponding field  $\hat{f}_e$  at a given single measurement  $\hat{v}$
- We assume that the field is smooth. Therefore, we smooth it to obtain an estimated field based on all of the measurements in a neighborhood location  $x$
- Correct the original image  $\hat{v}_{new}(\mathbf{x}) = \hat{v}_{old}(\mathbf{x}) - \hat{f}_s(\hat{v}(\mathbf{x}))$

Fields, 08, HmZhu

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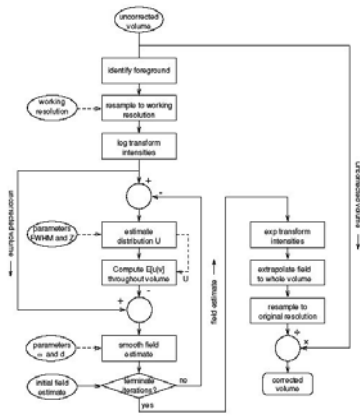
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### N3 Algorithm




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## 3D volume example 1

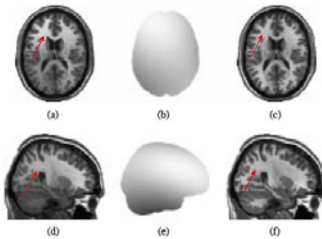


Fig. 12 Intensity nonuniformity correction of a T1 weighted 27-scan averaged gradient-echo MR scan: (a) and (d) transaxial and sagittal views of uncorrected data; (b) and (e) nonuniformity field estimated by the N3 method; (c) and (f) corrected data.

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### 3D volume example 2

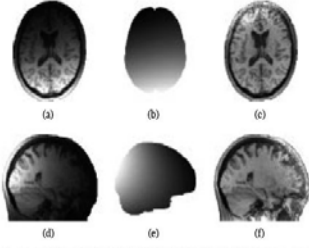


Fig 13 Intensity nonuniformity correction of a surface coil MR scan: (a) and (d) transaxial and sagittal views of uncorrected data; (b) and (e) nonuniformity field estimated by the N3 method; (c) and (f) corrected data.

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